



Turkish National Meeting on Automatic Control  
(TOK 2013) , Sept. 25, 2013, Malatya, Turkey



# A Tutorial on Fractional Order Motion Control

Part VI: Effects of FO Controls on Nonlinearities

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**MESA Lab** <http://mechatronics.ucmerced.edu>

- Fractional Order Motion Controls

John Wiley & Sons, Inc.

Hardcover, 454 pages, December 2012

- Dr. Ying Luo

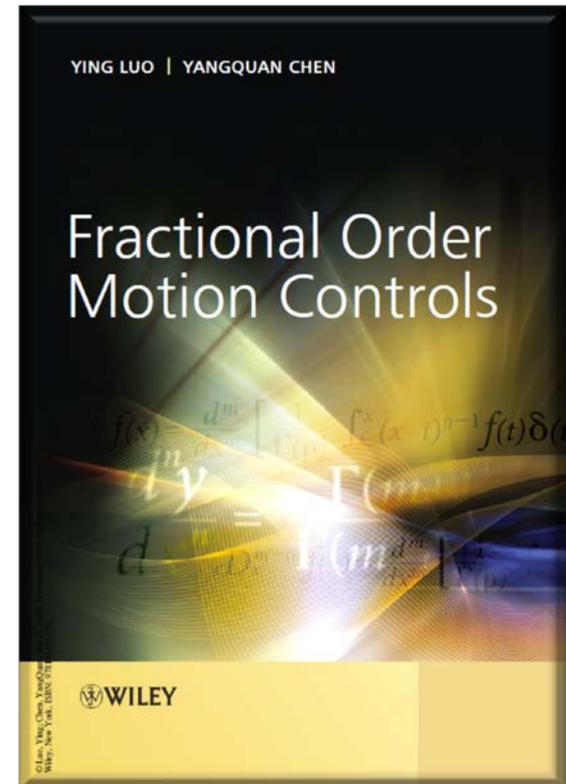
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Proceedings of the 2006 American Control Conference  
Minneapolis, Minnesota, USA, June 14-16, 2006

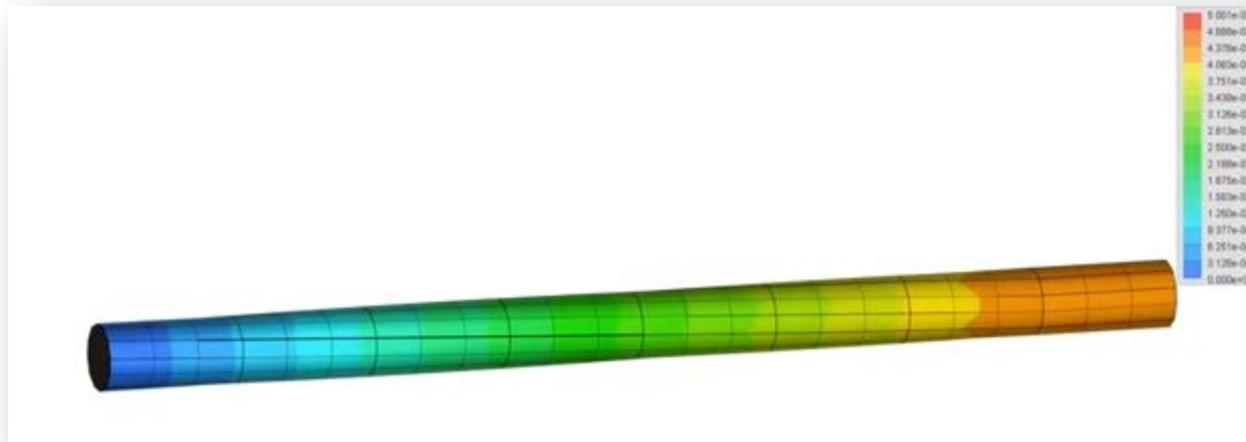
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## Fractional Order PID Control of A DC-Motor with Elastic Shaft: A Case Study

Dingyü Xue and Chunna Zhao  
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A video for Torsion (circular) test

<http://www.youtube.com/watch?v=76dVeHSWDZk>

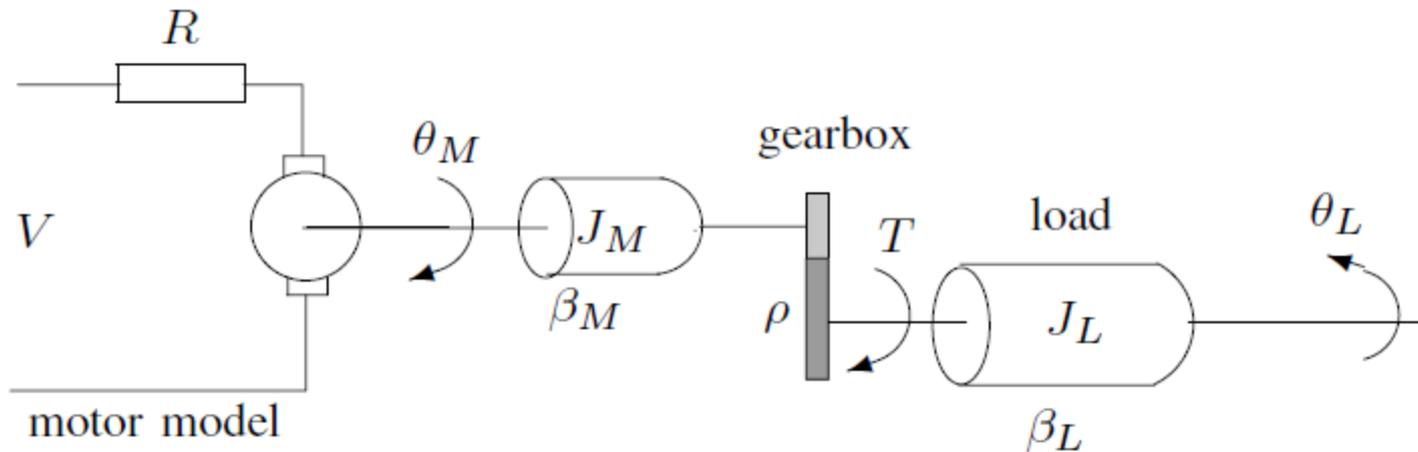


Figure: The benchmark position servomechanism model

### The plant model

$$\dot{\omega}_L = -\frac{k_\theta}{J_L} \left( \theta_L - \frac{\theta_M}{\rho} \right) - \frac{\beta_L}{J_L} \omega_L$$

$$\dot{\omega}_M = \frac{k_T}{J_M} \left( \frac{V - k_T \omega_M}{R} \right) - \frac{\beta_M \omega_M}{J_M} + \frac{k_\theta}{\rho J_M} \left( \theta_L - \frac{\theta_M}{\rho} \right)$$

## The plant model

$$\dot{x}_p = \begin{bmatrix} 0 & 1 & 0 & 0 \\ -\frac{k_\theta}{J_L} & -\frac{\beta_L}{J_L} & \frac{k_\theta}{\rho J_L} & 0 \\ 0 & 0 & 0 & 1 \\ \frac{k_\theta}{\rho J_M} & 0 & -\frac{k_\theta}{\rho^2 J_M} & -\frac{\beta_M + k_T^2/R}{J_M} \end{bmatrix} x_p + \begin{bmatrix} 0 \\ 0 \\ 0 \\ \frac{k_T}{R J_M} \end{bmatrix} V,$$

$$\theta_L = [1 \ 0 \ 0 \ 0]x_p,$$

$$T = \left[ k_\theta \ 0 -\frac{k_\theta}{\rho} \ 0 \right] x_p.$$

An SISO system

The input:  $V$

The Feedback:  $\theta_L$

Table: The system parameters of the servomechanism

Symbol	Value (SI units)	Symbol	Value (SI units)
$k_\theta$	1280.2	$\rho$	20
$k_T$	10	$\beta_M$	0.1
$J_M$	0.5	$\beta_L$	25
$J_L$	$50J_M$	$R$	20

## A proposed approximate realization method of fractional derivative

$$s^\alpha \approx \left( \frac{1 + \frac{s}{\frac{d}{b}\omega_b}}{1 + \frac{s}{\frac{b}{d}\omega_h}} \right)^\alpha$$

$$s^\alpha \approx K \left( \frac{ds^2 + bs\omega_h}{d(1-\alpha)s^2 + bs\omega_h + d\alpha} \right) \prod_{k=-N}^N \frac{s + \omega'_k}{s + \omega_k}$$

Better than Oustaloup's fractional operator approximation method

## The IOPID

Using ITAE

$$G_{c1} = 41.94 + \frac{21.13}{s} - 8.26s$$

Using ISE

$$G_{c1} = 110.09 + \frac{10.65}{s} - 30.97s$$

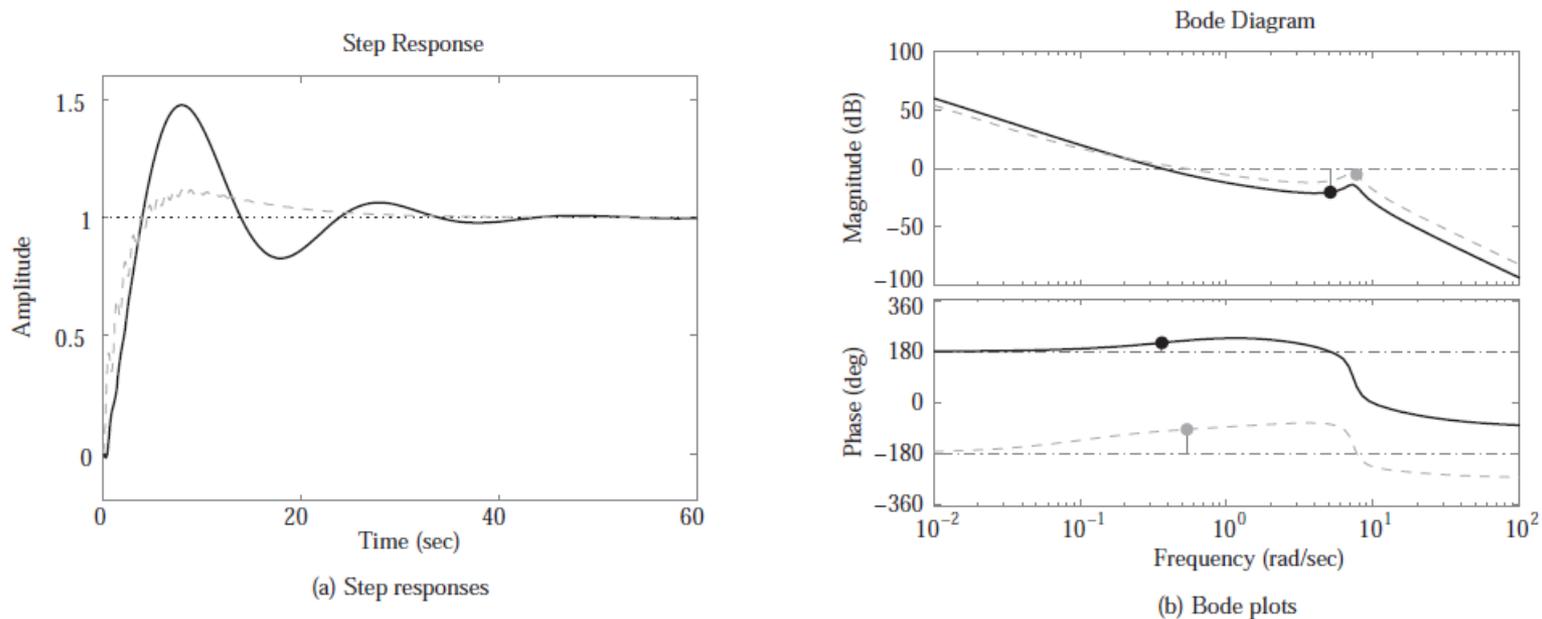


Figure: the best FOPID Controllers. Solid line: ITAE; Broken line: ISE

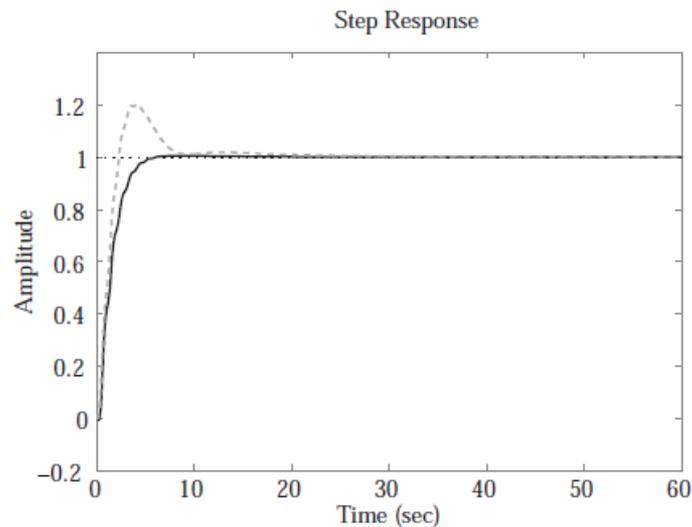
The FOPID with pre fixed  $\lambda = 0.5$  and  $\mu = 0.6$

Using ITAE

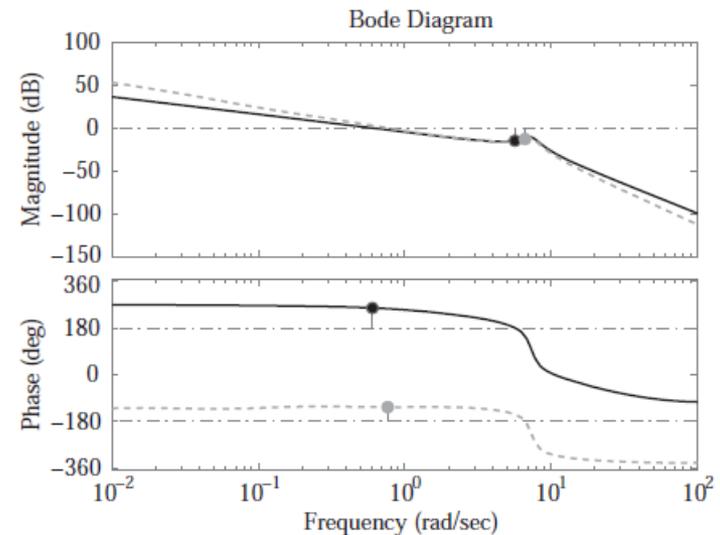
$$G_{c3} = 135.12 + \frac{0.01}{s^{0.5}} - 31.6s^{0.6}$$

Using ISE

$$G_{c4} = 61.57 + \frac{91.95}{s^{0.5}} - 2.33s^{0.6}$$

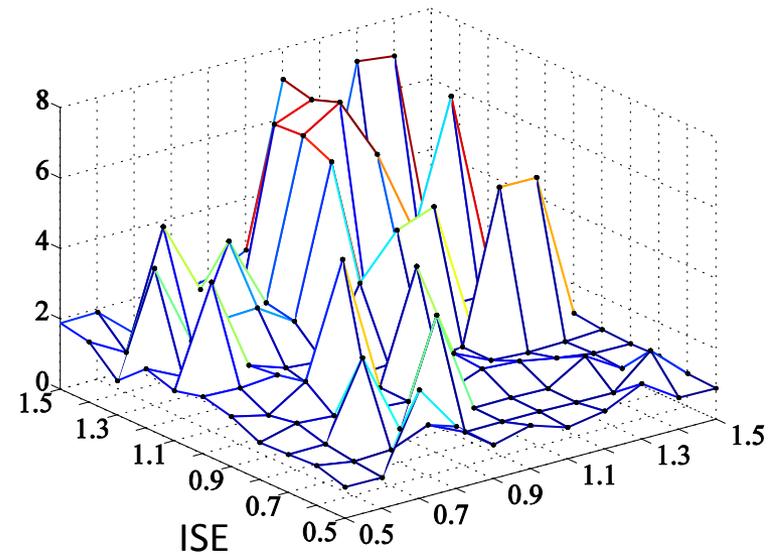
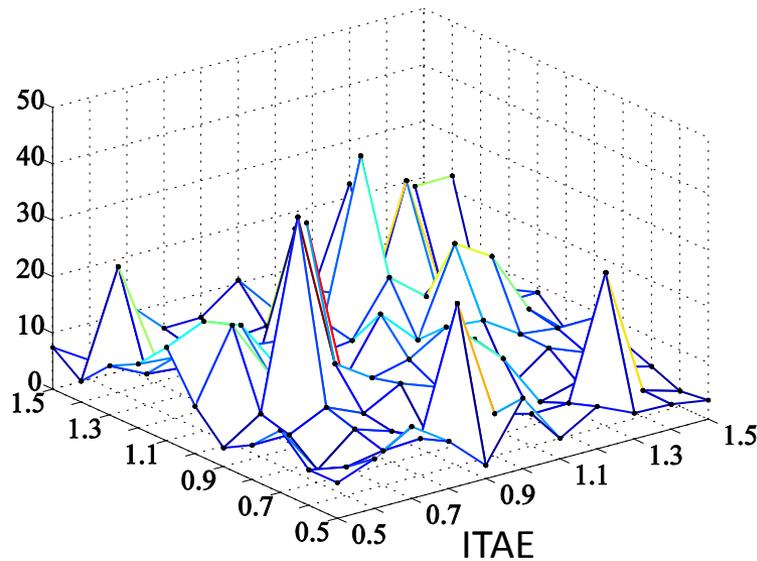


(a) Step responses



(b) Bode plots

Figure: the best IOPID Controllers. Solid line: ITAE; Broken line: ISE

Searching for the best fractional orders  $\lambda$  and  $\mu$  $N = 2, 4, 6, \dots$ Figure: Searching for the best fractional orders ( $N = 4$ )

## Robustness to load change

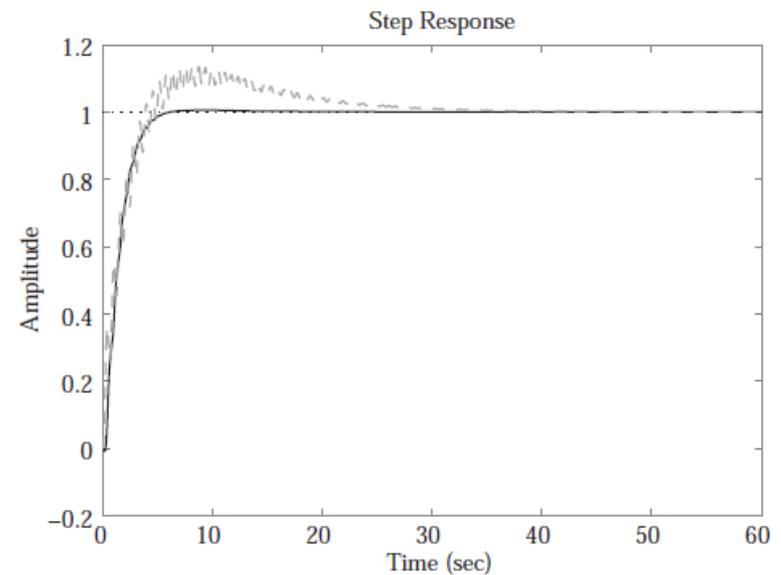
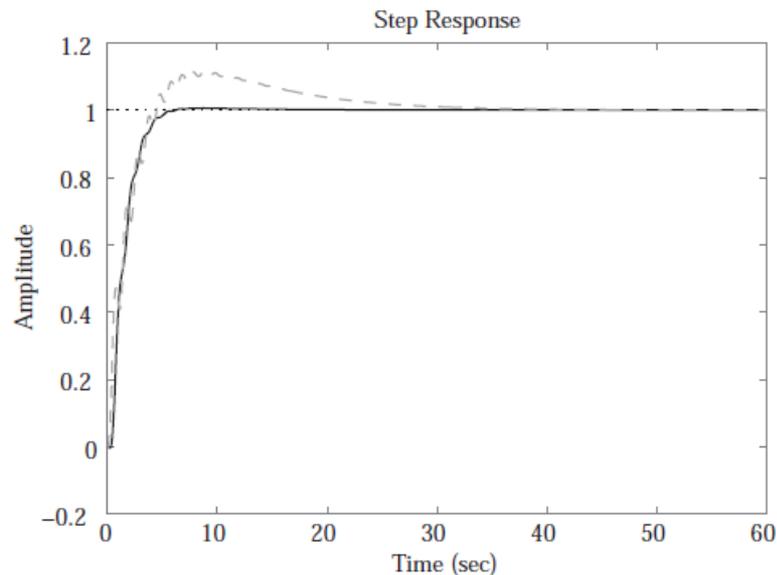
 $\pm 50\%$  load variation

Figure: Step responses comparison with  $N = 4$  (Dotted: Best PID; Solid line: Best FOPID)

## Robustness to Elasticity parameter change

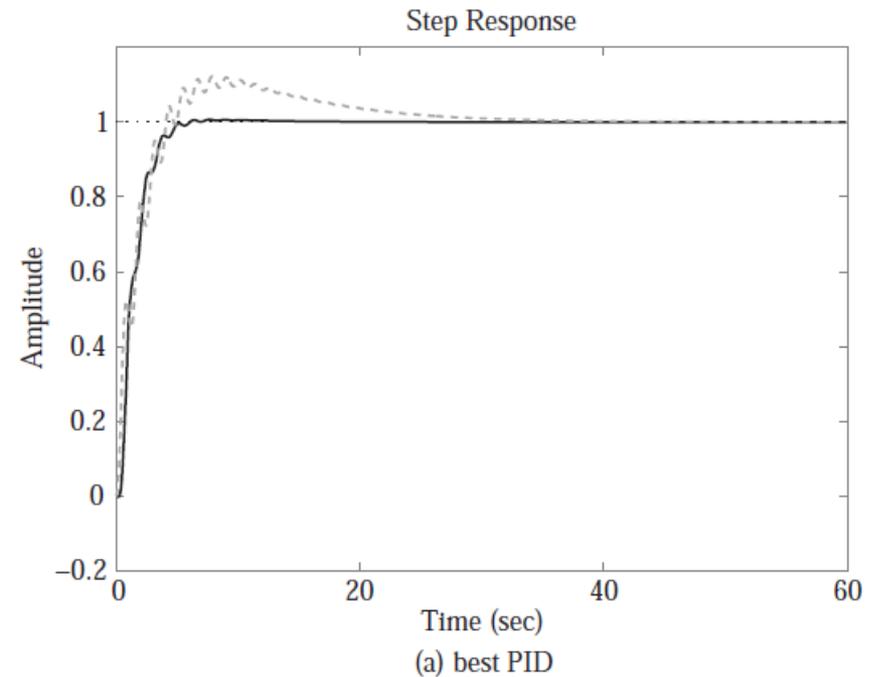
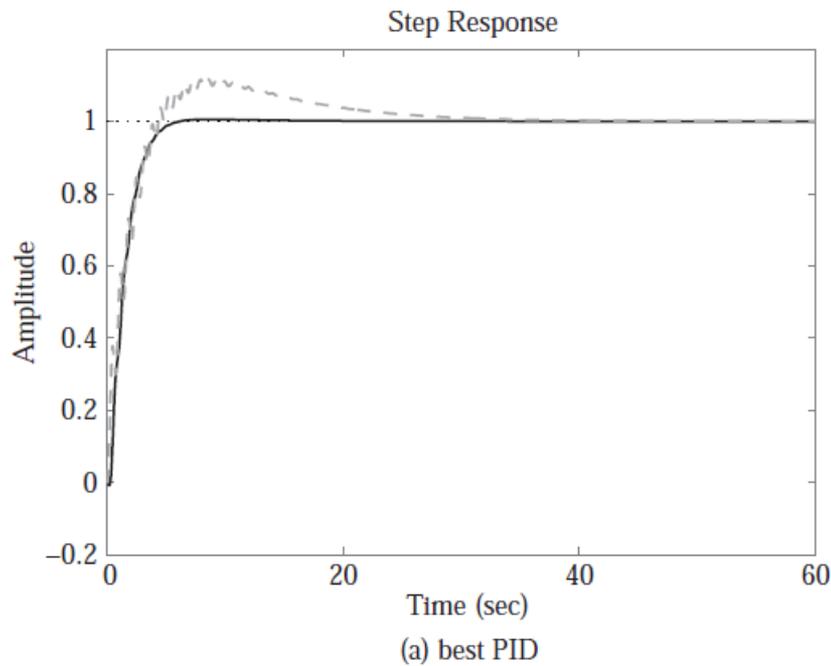
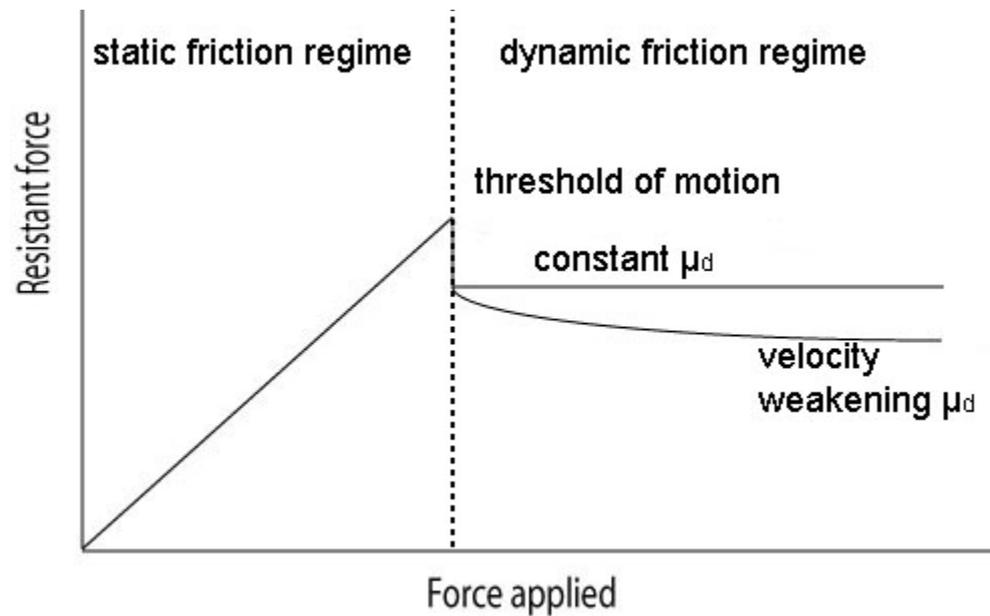
 $\pm 50\%$   $k_\theta$  load variation

Figure: Responses comparison when  $k_\theta$  increases 50% Dashed line: Best IO Controllers; Solid line: Best FO Controllers

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## The impact of friction force

Tracking error  
Limit cycle  
Spiky performance  
Etc.



Recall the plant model and the controller in Chapter 6

Second order position servo

$$P(s) = \frac{1.52}{s(0.4s + 1)}$$

Friction was not considered back to ch6

Now, take it into consideration below

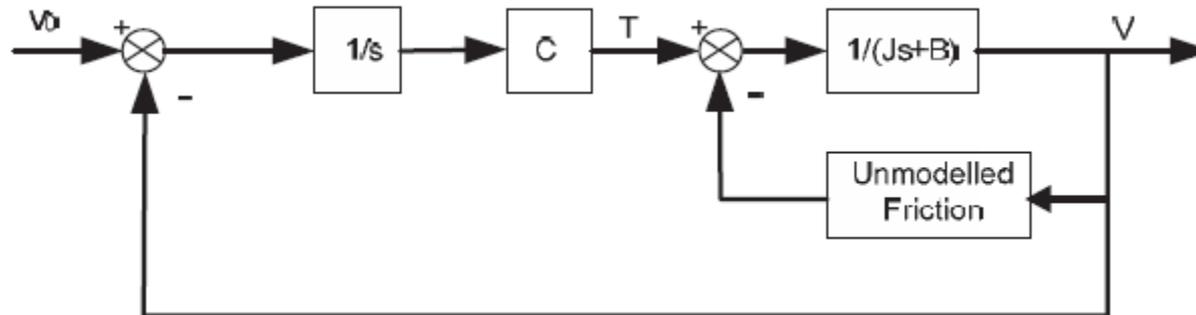


Figure: Position tracking control equivalent diagram with constant speed reference and friction

## Notice

Ultra-low speed !

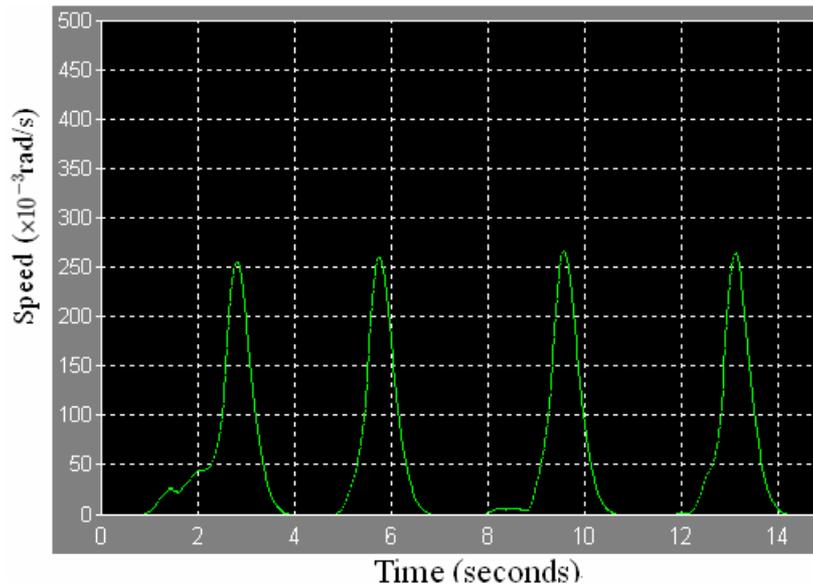
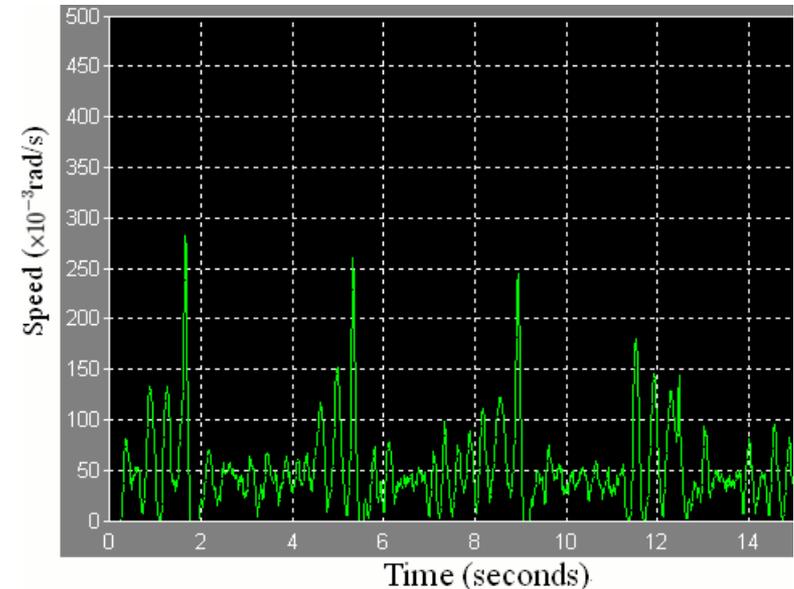
IOPI:  $C(s)=2.56(1+1.166/s)$ FOPD:  $C(s)=13.86(1+0.368/s)$ 

Figure: Speed outputs of position tracking with constant speed reference using optimized IOPI and designed FOPD

# A New Model for Control of Systems with Friction

C. Canudas de Wit, *Associate, IEEE*, H. Olsson, *Student Member, IEEE*, K. J. Åström, *Fellow, IEEE*, and P. Lischinsky

$$\begin{aligned}\dot{z} &= v - \frac{|v|}{g(v)}z, \\ F &= \sigma_0 z + \sigma_1 \dot{z} + \sigma_2 v, \\ \sigma_0 g(v) &= F_C + (F_S - F_C)e^{-(v/v_s)^2},\end{aligned}$$

Figure: Parameter values in the LuGre model

$\sigma_0$	$10^5$	[N/m]
$\sigma_1$	$\sqrt{10^5}$	[Ns/m]
$\sigma_2$	0.4	[Ns/m]
$F_C$	1	[N]
$F_S$	1.5	[N]
$v_s$	0.001	[m/s]

where the average deflection of the bristles is denoted by  $z$ ;  $v$  is the relative velocity between the two surfaces; the function  $g$  is positive and depends on many factors such as material properties, lubrication and temperature;  $\sigma_0$  is the stiffness, and  $\sigma_1$ ,  $\sigma_2$  are damping coefficients;  $F_C$  is the Coulomb friction level;  $F_S$  is the level of the stiction force; and  $v_s$  is the Stribeck velocity. The values of those parameters in (18.1)

## The describing function method

The ratio of the first harmonic of output to input

$$y_1(t) = a \cos \omega t + b \sin(\omega t) = c \sin(\omega t + \varphi),$$

where

$$a = \frac{2}{\pi} \int_0^{\pi} y(t) \cos(\omega t) d(\omega t), \quad b = \frac{2}{\pi} \int_0^{\pi} y(t) \sin(\omega t) d(\omega t),$$

$$c = \sqrt{a^2 + b^2}, \quad \varphi = \arctan(a/b).$$

## Decoupling the linear and nonlinear part

$$N(T_0, A, \omega) = [|N|e^{\phi}, v_a]$$

$T_0$  is the mean torque

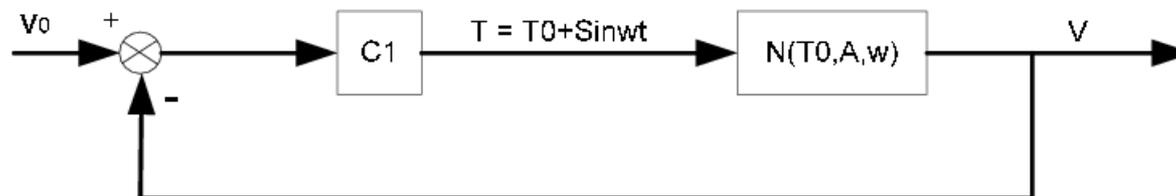
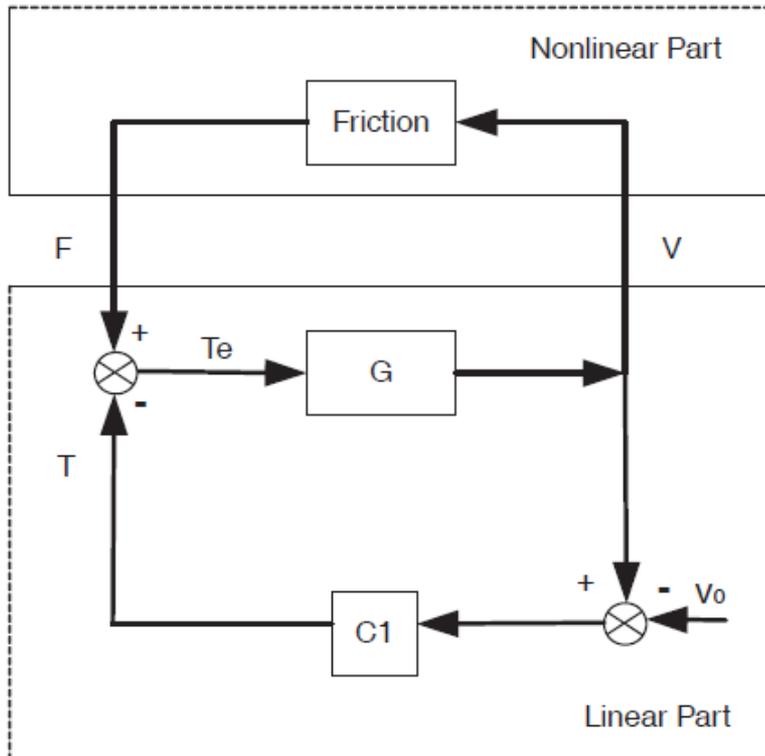
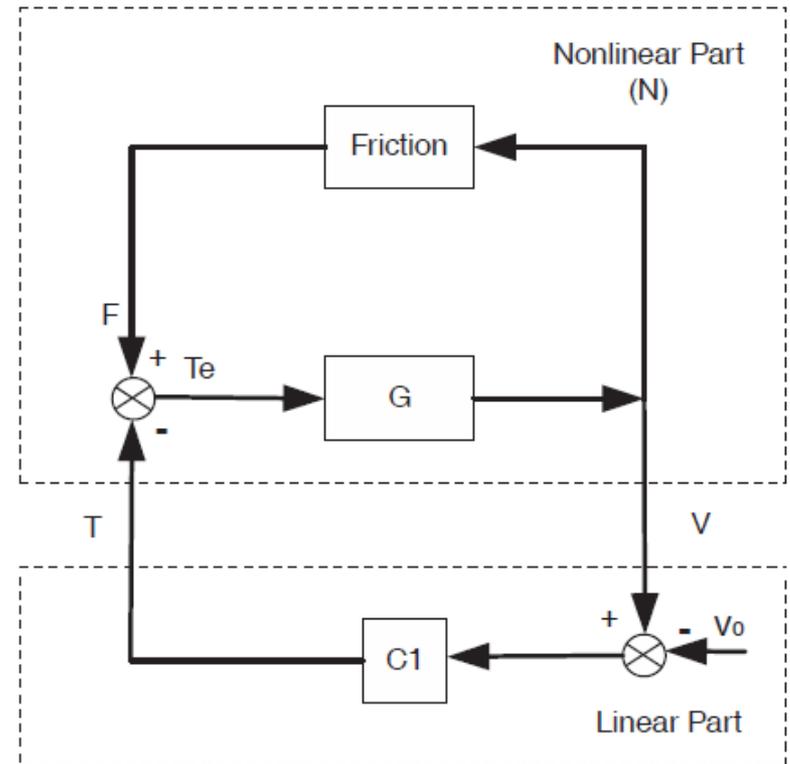


Figure: Approximation closed-loop system with linear and nonlinear parts



(a) Nonlinear part with only friction



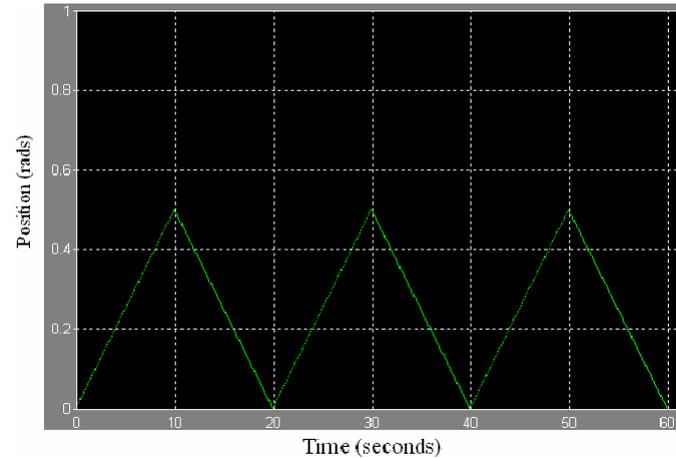
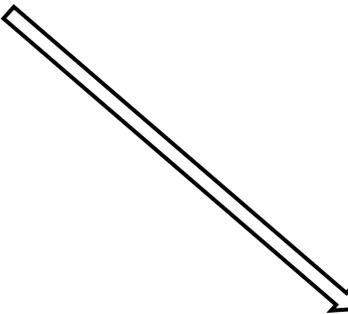
(b) Nonlinear part with friction and system dynamics

Figure: Two methods of uncoupling linear and nonlinear parts

Ultra-low speed reference

FOPD:  $C(s)=2.56(1+1.166/s)$

IOPI:  $C(s)=13.86(1+0.368/s)$



Varying low speed ( $\pm 0.05\text{rad/s}$ ) position reference for tracking

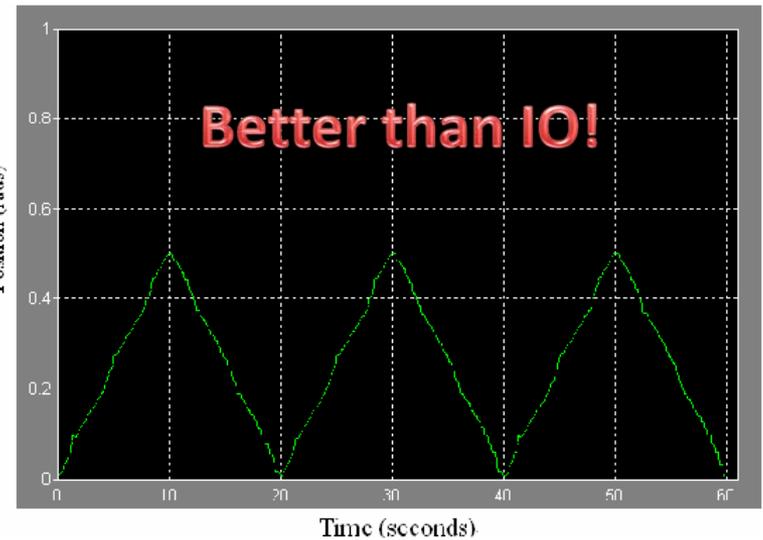
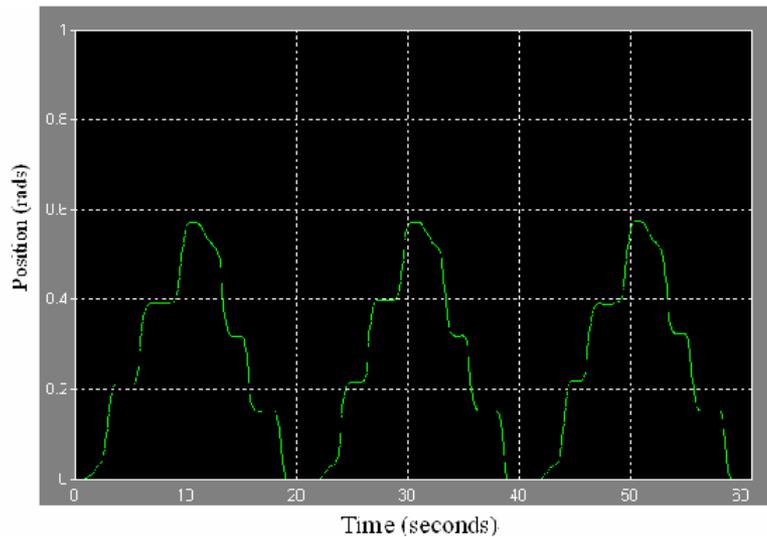


Figure: Position tracking outputs with varying low speed reference using IOPI/FOPD

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### The integrator: a double-edged sword

Eliminating the steady state error  
Introducing a  $90^\circ$  phase lag

### Attempts of improvement

CCI: Clegg conditional integrator  
ICI: intelligent conditional integrator  
MICI: modified intelligent conditional integrator  
FOCI: fractional order conditional integrator  
OFOCI: optimized fractional order conditional integrator

# A Nonlinear Integrator for Servomechanisms

J. C. CLEGG  
ASSOCIATE MEMBER AIEE

MARCH 1958

CCI

« Information and Control » 1990-06

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## A NEW INTELLIGENT INTEGRATOR

WANG Yong CAO Zhengping XIANG Guobo (Dept of Automation, Wuhan University of Technology)

ICI

## The input and output of the CCI

For input  $x(t)A\sin(\omega t)$

$$\begin{aligned} y_C(\varphi) &= \int_0^{\varphi/\omega} x(t)dt = \frac{1}{\omega} \int_0^{\varphi} A\sin(\omega t)d(\omega t) \\ &= \frac{A}{\omega}(1 - \cos(\varphi)). \end{aligned}$$

## The describing function of the CCI

$$\begin{aligned} N_C(A, \omega) &= \frac{2j}{\pi A} \int_0^{\pi} \frac{A}{\omega}(1 - \cos \varphi)e^{-j\varphi}d\varphi = \frac{4}{\pi\omega} \left(1 - j\frac{\pi}{4}\right), \\ |N_C(A, \omega)| &= \frac{1}{\omega} \sqrt{1 + \left(\frac{4}{\pi}\right)^2} = \frac{1.62}{\omega}, \\ \arg N_C(A, \omega) &= -\arctan \frac{\pi}{4} \simeq 0.212\pi(\text{rad.}) = 38.15^\circ. \end{aligned}$$

## The input and output of the FOCI

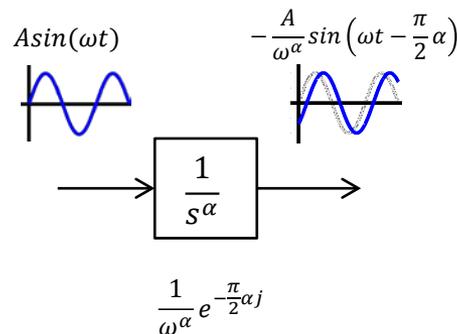
The FO integrator

$${}_{-\infty}D_t^{-\alpha} A \sin(\omega t) = \frac{A}{\omega^\alpha} \sin\left(\omega t - \frac{\alpha\pi}{2}\right),$$

The output of the FOCI

$$y_F(t) = \begin{cases} {}_0D_t^\alpha u(t) = \frac{A}{\omega^\alpha} [\sin(\omega t - \frac{\alpha\pi}{2}) + \sin(\frac{\alpha\pi}{2})], & 0 \leq \omega t < \frac{\pi}{2}; \\ k \frac{A}{\omega}, \frac{\pi}{2} \leq \omega t < \pi; \\ 0, & \omega t = \pi; \\ {}_\pi D_t^\alpha u(t) = \frac{A}{\omega^\alpha} [\sin(\omega t - \frac{\alpha\pi}{2}) - \sin(\frac{\alpha\pi}{2})], & \pi \leq \omega t < \frac{3\pi}{2}; \\ -k \frac{A}{\omega}, \frac{3\pi}{2} \leq \omega t < 2\pi; \\ 0, & \omega t = 2\pi; \end{cases}$$

k is the “holding” parameter



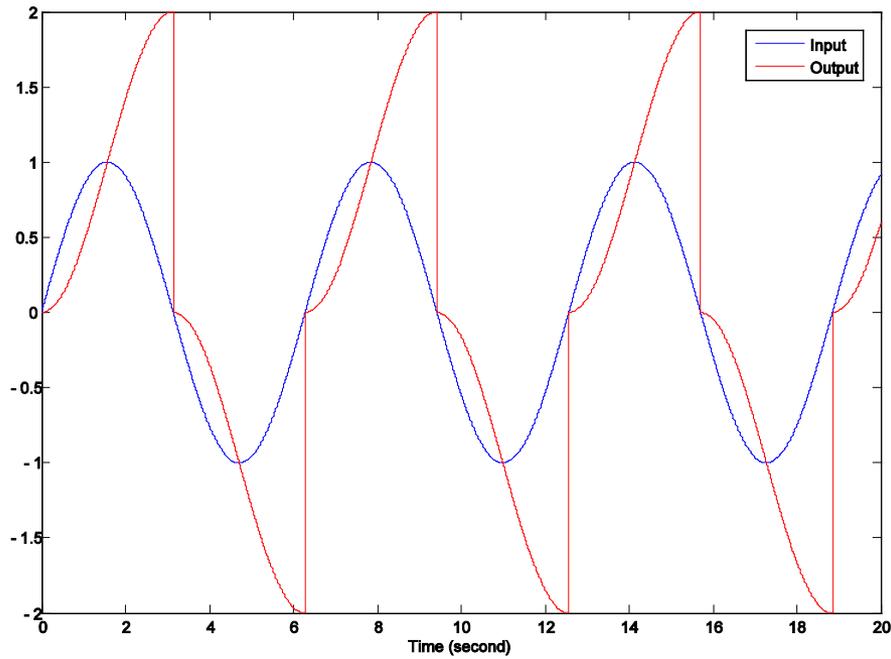


Figure: Input and output signals of CCI

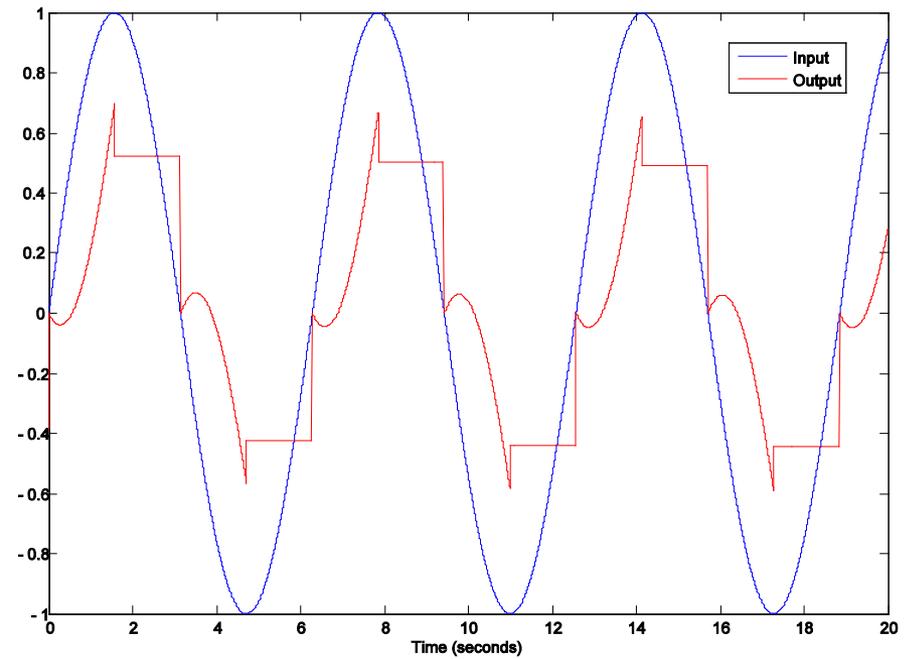


Figure: Input and output signals of FOCI

## Optimality criteria

- $\text{Min} \frac{\partial J_{xF}(\alpha, k)}{\partial \alpha} = 0$   
 $J_{xF}(\alpha, k) = \frac{C_{xF}}{C_{1F}}$  is the cost function,  $C_{xF}$  is the x-th order harmonic
- Phase delay  
 $\varphi_{1F} = \arctan\left(\frac{a_{1F}}{b_{1F}}\right) = \varphi^*$

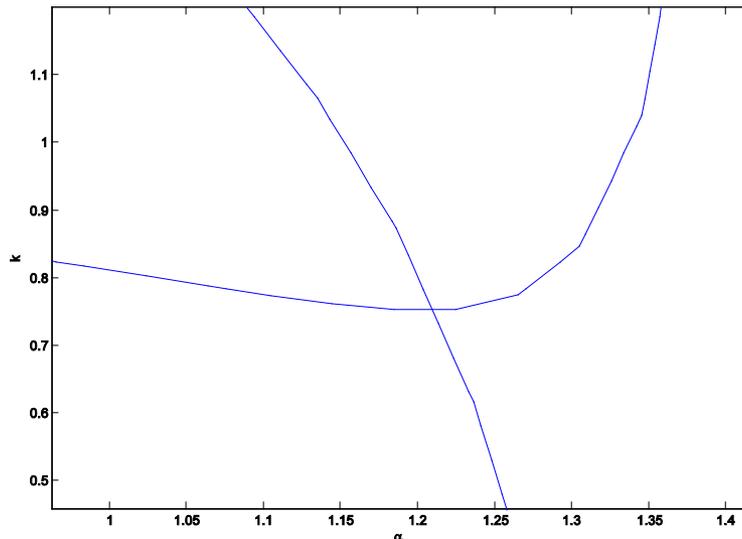


Figure: Optimized solutions with phase delays  $-32^\circ$  according to the optimality criteria

Table: Performance comparison with the third-order harmonic component ratio minimum specification

Type	$N(j\omega)$	$\varphi_1$	$\frac{c_3}{c_1}$ (%)	$\frac{c_5}{c_1}$ (%)	$\frac{c_7}{c_1}$ (%)	$\frac{c_9}{c_1}$ (%)	$\Sigma$ (%)
<i>CCI</i>	$\frac{1.62}{\omega} e^{-j0.212\pi}$	$-38.1^\circ$	26.2	15.8	11.2	8.7	61.9
<i>ICI</i>	$\frac{1.08}{\omega} e^{-j0.153\pi}$	$-27.6^\circ$	9.84	13.78	7.03	7.22	37.87
<i>OFOCI<sub>1</sub></i>	$\frac{0.42}{\omega^{1.273}} e^{-j0.212\pi}$	$-38.1^\circ$	6.23	17.35	5.24	8.51	37.33
<i>OFOCI<sub>2</sub></i>	$\frac{0.52}{\omega^{1.209}} e^{-j0.178\pi}$	$-32.0^\circ$	0.31	16.29	5.77	8.14	30.51
<i>OFOCI<sub>3</sub></i>	$\frac{0.67}{\omega^{1.126}} e^{-j0.153\pi}$	$-27.6^\circ$	4.38	15.04	6.25	7.65	33.32
<i>OFOCI<sub>4</sub></i>	$\frac{1.33}{\omega^{0.530}} e^{-j0.122\pi}$	$-22.0^\circ$	12.48	12.73	7.10	6.75	39.06

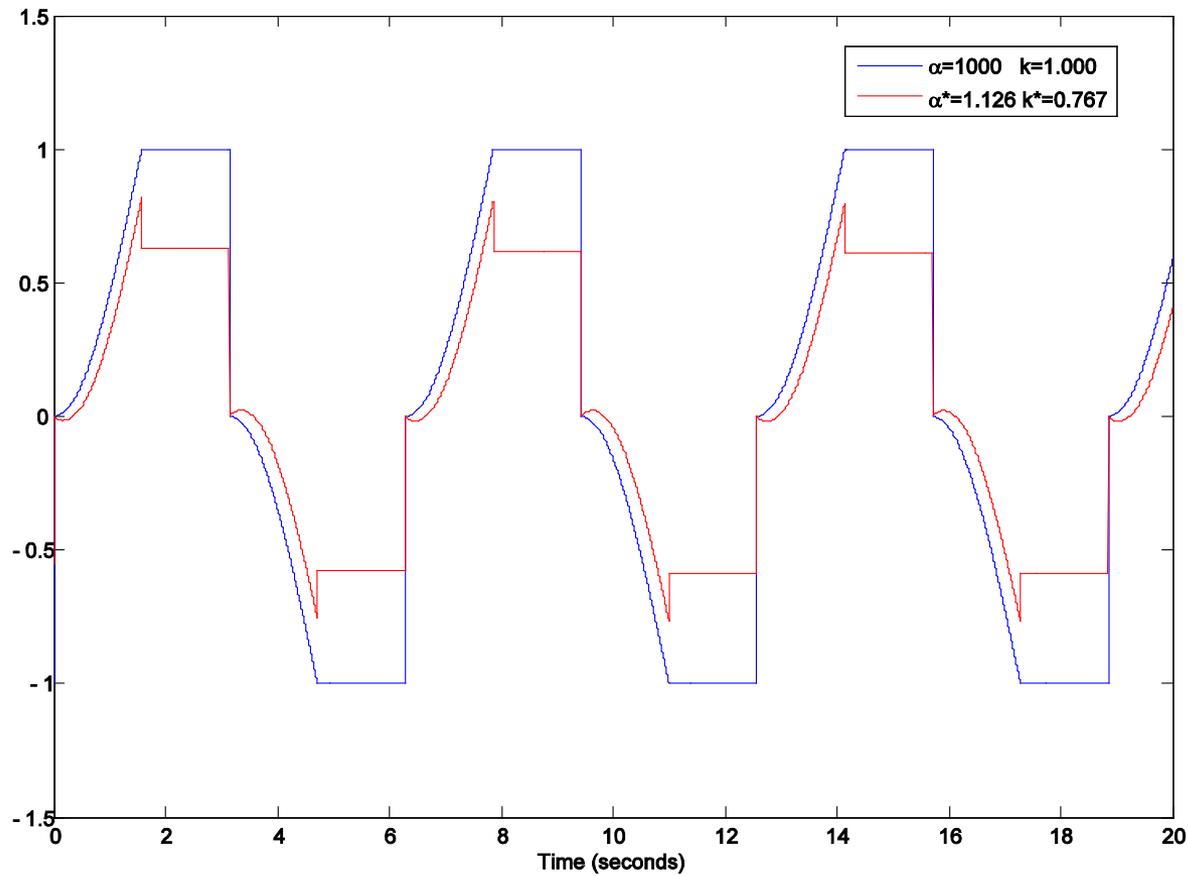


Figure: Output comparison of the ICI and the OFOCI3

**This is the end of session VI**

**Questions?**